

ECE 6604 Homework #5 Solutions

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1. (3) Compare the *peak-to-mean envelope power ratio* (PMEPR) of QPSK and $\pi/4$ -DQPSK.

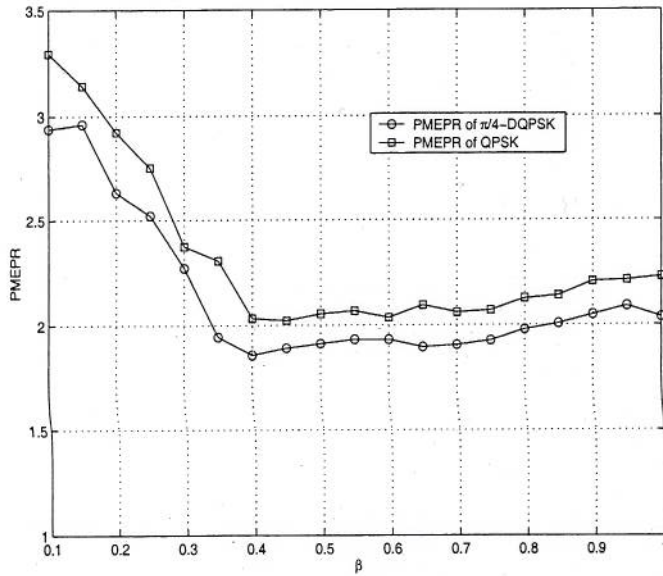


Figure 2: PMEPR Performance of $\pi/4$ -DQPSK and QPSK

The *peak-to-mean envelope power ratio* of $\pi/4$ -DQPSK is lower than that of QPSK. This indicates that $\pi/4$ -DQPSK is less sensitive to the amplifier non-linearity than QPSK.

2: a) Note that $h_f(t)$ should integrate to $1/2$. So the peak value of $h_f(t)$ should have been $1/T$ and not $1/2T$. This would give a peak frequency deviation of $h \times \frac{1}{2} \times (M-1) = (M-1)/2T$.

b)

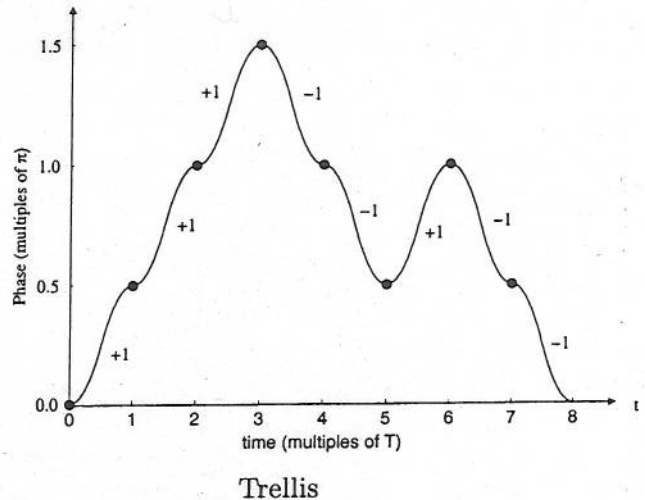
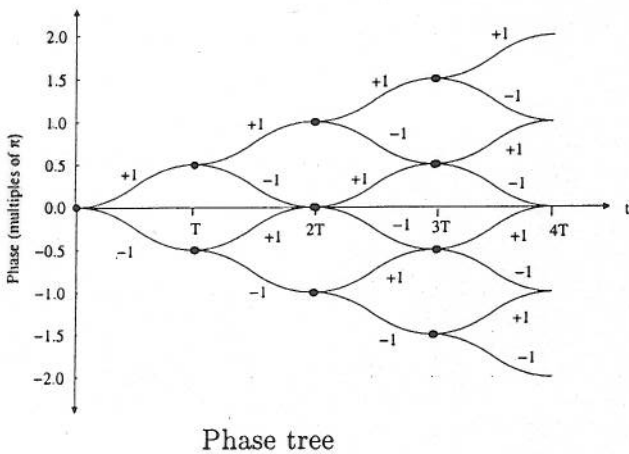


Figure 1: Phase tree and trellis for the sequence $\mathbf{x} = (+1, +1, +1, -1, -1, +1, -1, -1)$.

A Gaussian pulse-shaping filter has the transfer function

$$H(f) = \exp \left\{ -\left(\frac{f}{B}\right)^2 \frac{\ln 2}{2} \right\}.$$

The impulse response of the filter is

$$\begin{aligned} h(t) &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} H(f) \exp j2\pi f t df \\ &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} \exp \left\{ -\left(\frac{f}{B}\right)^2 \frac{\ln 2}{2} + j2\pi f t \right\} df \\ &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} \exp \left\{ -\frac{1}{B^2} \frac{\ln 2}{2} \left[f - j2\pi t \frac{B^2}{\ln 2} \right]^2 \right\} \cdot \exp \left\{ -2\pi^2 t^2 \frac{B^2}{\ln 2} \right\} df \\ &= \frac{1}{2\pi} \exp \left\{ -2\pi^2 t^2 \frac{B^2}{\ln 2} \right\} \int_{-\infty}^{+\infty} \exp \left\{ -\frac{1}{B^2} \frac{\ln 2}{2} f^2 \right\} \cdot df \\ &= \frac{1}{\sqrt{2\pi}} \exp \left\{ -2\pi^2 t^2 \frac{B^2}{\ln 2} \right\} \cdot B \sqrt{\frac{1}{\ln 2}} \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{+\infty} \exp \left\{ -\frac{1}{2} s^2 \right\} \cdot ds \\ &= \frac{1}{\sqrt{2\pi \ln 2}} B \exp \left\{ -2\pi^2 t^2 \frac{B^2}{\ln 2} \right\}. \end{aligned}$$

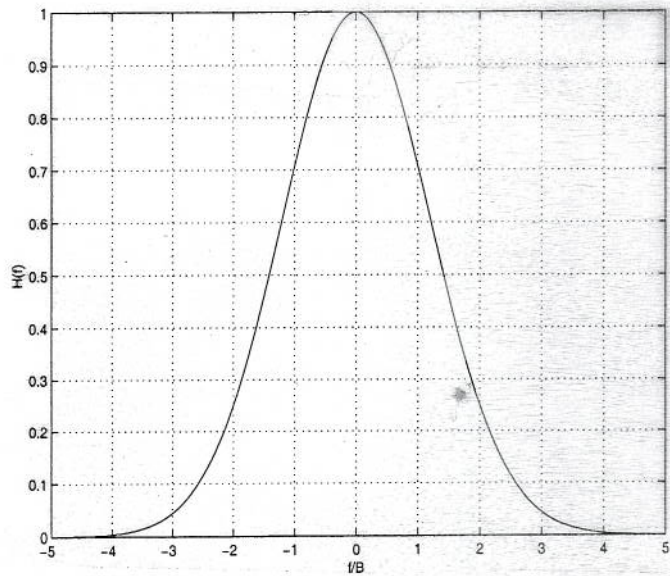
The frequency shaping pulse is

$$h_f(t) = \frac{1}{2T} \left[Q \left(\frac{t/T + 1/2}{\sigma} \right) - Q \left(\frac{t/T - 1/2}{\sigma} \right) \right],$$

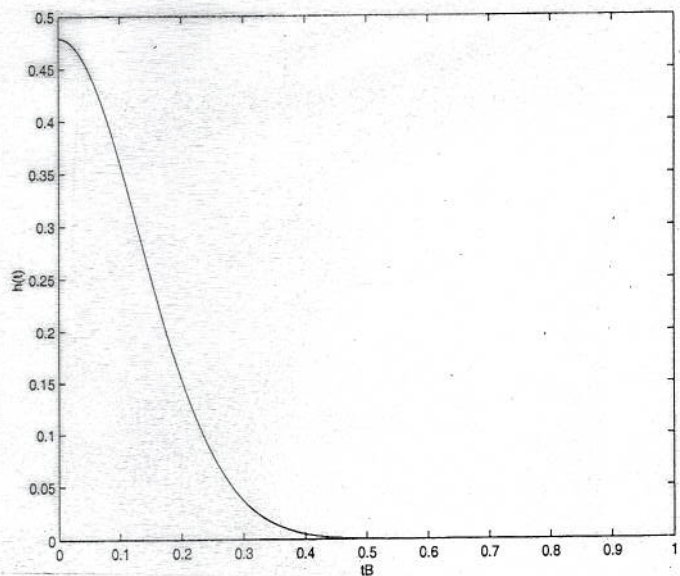
where

$$\sigma^2 = \frac{\ln 2}{4\pi^2 (BT)^2}.$$

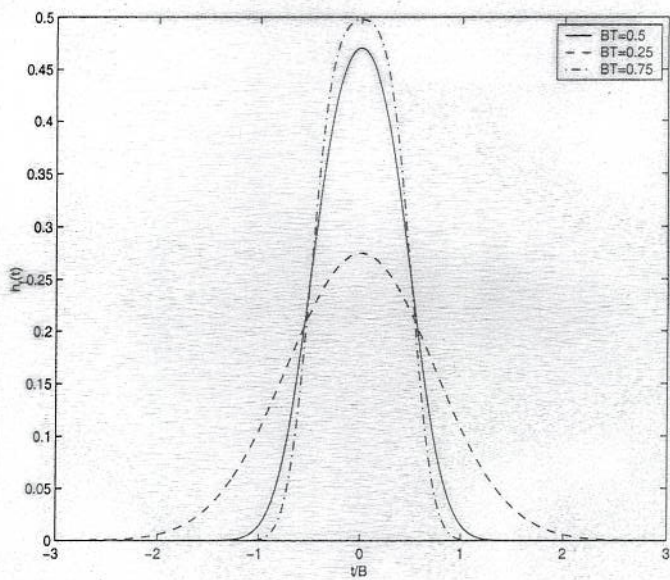
The frequency response and the impulse response of the filter, and the frequency shaping pulse $h_f(t)$ are shown in the respective figures.



Frequency Response $H(f)$



Impulse Response $h(t)$



Frequency Shaping Pulse $h_f(t)$

18 a) We need to calculate the autocorrelation function of the binary sequence \mathbf{x} , $x_i \in \{-1, +1\}$ such that $P(x_n = x_{n+1}) = 3/4$ and $P(x_n \neq x_{n+1}) = 1/4$. Consider $\phi_{xx}(0)$ and $\phi_{xx}(1)$, i.e.

$$\begin{aligned} \phi_{xx}(0) &= E[x_n x_n] = 1 \text{ and} \\ \phi_{xx}(1) &= E[x_n x_{n+1}] = (1) \cdot P(x_n = x_{n+1}) + (-1) \cdot P(x_n \neq x_{n+1}) = 3/4 - 1/4 = 1/2 \end{aligned}$$

We claim that the autocorrelation function of this data sequence is $\phi_{xx}(m) = (\frac{1}{2})^{|m|}$ and we will prove it by induction. We have already shown that $\phi_{xx}(1) = 1/2$. Assuming $\phi_{xx}(m) = (\frac{1}{2})^{|m|}$, we will show $\phi_{xx}(m+1) = (\frac{1}{2})^{|m+1|}$. For $m \geq 2$, observe that

$$\begin{aligned} \phi_{xx}(m+1) &= E[x_n x_{n+m+1}] \\ &= E[x_{n+1} x_{n+m+1}] \cdot P(x_n = x_{n+1}) + -E[x_{n+1} x_{n+m+1}] \cdot P(x_n \neq x_{n+1}) \\ &= \phi_{xx}(m) \cdot \frac{3}{4} - \phi_{xx}(m) \cdot \frac{1}{4} \\ &= \frac{1}{2} \phi_{xx}(m) = \left(\frac{1}{2}\right)^{m+1} \end{aligned}$$

Similarly, we can show that $\phi_{xx}(m) = (\frac{1}{2})^{-m}$ for $m \leq 0$.

b) The power spectrum $S_{xx}(f)$ of the binary data sequence \mathbf{x} is

$$\begin{aligned} S_{xx}(f) &= \sum_{m=-\infty}^{\infty} \phi_{xx}(m) e^{-j2\pi f m} = \sum_{m=-\infty}^{\infty} \left(\frac{1}{2}\right)^{|m|} e^{-j2\pi f m} \\ &= -1 + \sum_{m=0}^{\infty} \left(\frac{1}{2}\right)^m e^{-j2\pi f m} + \sum_{m=0}^{\infty} \left(\frac{1}{2}\right)^m e^{j2\pi f m} \\ &= -1 + \frac{1}{1 - \frac{1}{2} e^{-j2\pi f}} + \frac{1}{1 - \frac{1}{2} e^{j2\pi f}} \\ &= \frac{2 - \cos(2\pi f)}{5/4 - \cos(2\pi f)} - 1 = \frac{3}{5 - 4 \cos(2\pi f)}. \end{aligned}$$