

EE6604

Personal & Mobile Communications

Lecture 27

OFDM Decision-Aided Reconstruction
and Residual ISI Cancellation

Clipped Signal Prior to Transmission

- The N -point IFFT output sequence is

$$X_k = \frac{1}{\sqrt{N}} \sum_{n=0}^{N-1} x_n \exp \left\{ j \frac{2\pi nk}{N} \right\}, \quad 0 \leq k \leq N-1$$

- Suppose we clip the IFFT output sequence as

$$Y_k = \begin{cases} X_k & , \quad |X_k| \leq A \\ A \exp \{ \arg(X_k) \} & , \quad |X_k| > A \end{cases} \quad 0 \leq k \leq N-1$$

– Note that the amplitude is clipped but the phase is maintained.

- The clipping ratio (CR) is defined as

$$\text{CR} = 20 \log \frac{A}{\sigma} \text{ dB} \quad \sigma = \sqrt{\text{E} [|X_n|^2]}$$

where σ is the rms signal power of x_k .

- A guard interval is added as

$$Y_k^g = Y_{(k+N-G)_N}, \quad 0 \leq k \leq N+G-1$$

Demodulated Signal with Clipping

- The received samples after removal of the guard interval are

$$R_k = \sum_{m=0}^M h_m Y_{(k-m)_N} + n_k, \quad 0 \leq k \leq N - 1$$

where h_m is the channel impulse response at lag m .

- Performing an FFT on $\{R_k\}$ gives

$$\begin{aligned} Z_n &= \eta_n y_n + w_n \\ &= \alpha_n \eta_n x_n + q_n, \quad 0 \leq k \leq N - 1 \end{aligned}$$

where the complex channel gain is

$$\eta_n = \sum_{m=0}^M h_m \exp \left\{ -j \frac{2\pi n m}{N} \right\}$$

Decision-Aided Reconstruction

1. From the demodulated samples $\{Z_n\}_{n=0}^{N-1}$, an estimate of the clipped signal, $\{\hat{Y}_k\}$, is obtained and stored in memory by performing IFFT on $\{\hat{y}_n\}$, where

$$\hat{y}_n = \frac{Z_n}{\eta_n}, \quad 0 \leq n \leq N - 1$$

2. Decisions on the transmitted symbols are made in the frequency domain as

$$\hat{x}_n^{(I)} = \min_{\{x\}} |Z_n^{(I)} - \eta_n x|, \quad 0 \leq n \leq N - 1$$

where I represents an iteration number with an initial value of $I = 0$, and $Z_n^{(0)} = Z_n$.

3. The decisions in Step 2 are converted back to the time domain using an IFFT, yielding $\{\hat{X}_k^{(I)}\}$.
4. The clipped samples are detected and new sequence, $\{\hat{Y}_k^{(I)}\}$, is generated as

$$\hat{Y}_k^{(I)} = \begin{cases} \hat{Y}_k & , \quad \left| \hat{X}_k^{(I)} \right| \leq A \\ \hat{X}_k^{(I)} & , \quad \left| \hat{X}_k^{(I)} \right| > A \end{cases}, \quad 0 \leq k \leq N - 1$$

where the sequence $\{\hat{Y}_k\}$ is *always* retrieved from the memory in Step 1.

Decision-Aided Reconstruction (cont'd)

5 The sequence $\{\hat{Y}_k^{(I)}\}$ is converted to the frequency domain yielding $\{\hat{y}_n^{(I)}\}$.

6 Increment the index number $I \leftarrow I + 1$ and determine $\{Z_n^{(I)}\}$ by

$$Z_n^{(I)} = \eta_n \hat{y}_n^{(I-1)}, \quad 0 \leq n \leq N - 1 .$$

7 Go back to Step 2. Decisions are made yielding $\{\hat{x}_n^{(I)}\}$. This completes the I th iteration of the DAR algorithm.

8 Continue iterations by repeating Steps 3–7.

- When the clipping noise is large compared to the AWGN, the performance is limited by an error floor due to the clipping noise.
- The clipping noise is mitigated significantly when the decisions are made in the frequency domain; when the symbol decisions are converted back to the time domain, the OFDM signal samples that had a large amplitude and were clipped in the transmitter will actually re-grow. Although these samples will still be somewhat distorted by clipping, they are now much less affected by the clipping noise.

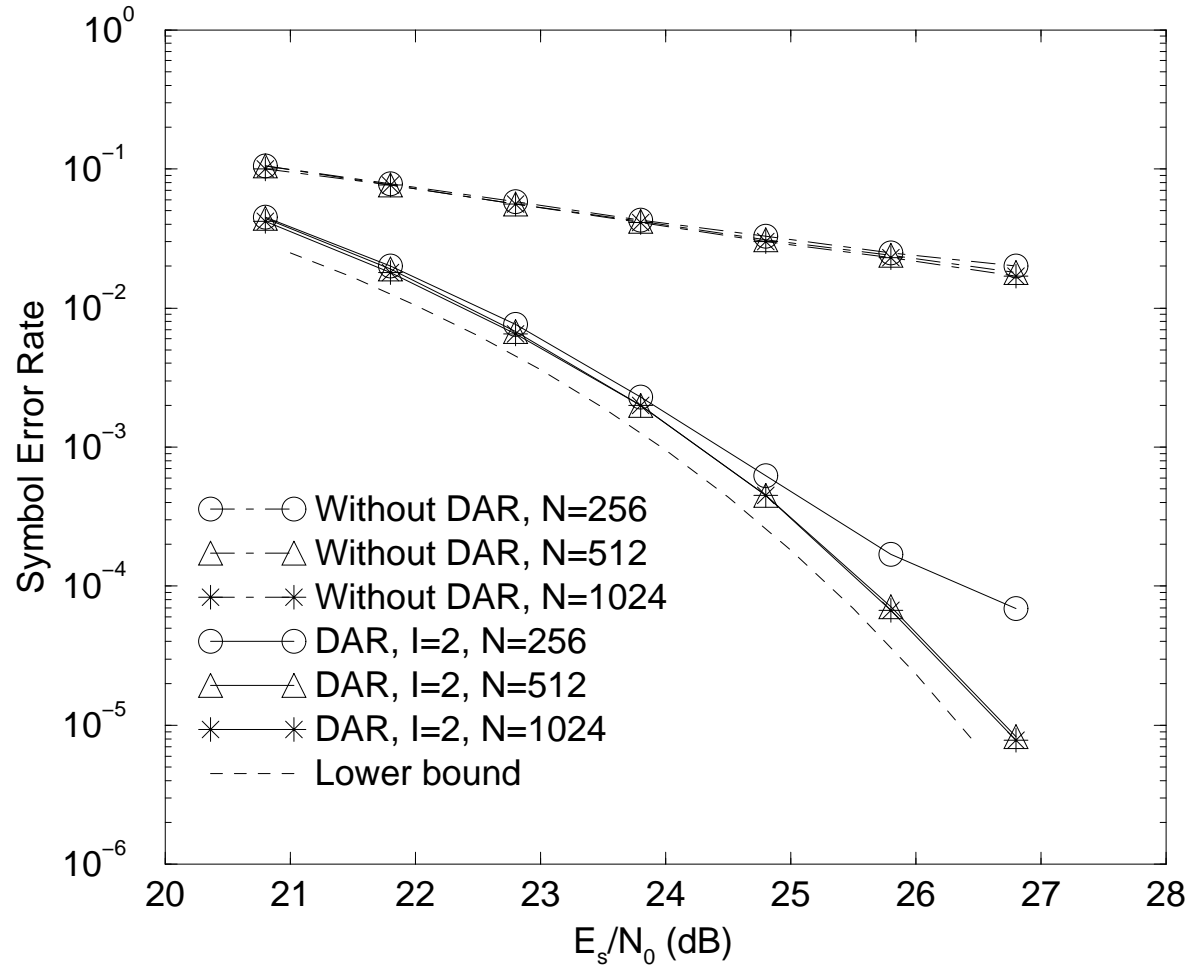
Simulations

- Consider using M -QAM with sample spacing $T_s^g = 0.2 \mu\text{s}$.
- The tap coefficients for the static ISI channel are listed below. This channel has several deep subchannel nulls, and the smallest subchannel power is -23.3 dB below the average subchannel power.
- A guard interval of length $10 \mu\text{s}$ is used, which is longer than the maximum excess delay of the channel.

Table 1: Tap Coefficients for the ISI Channel

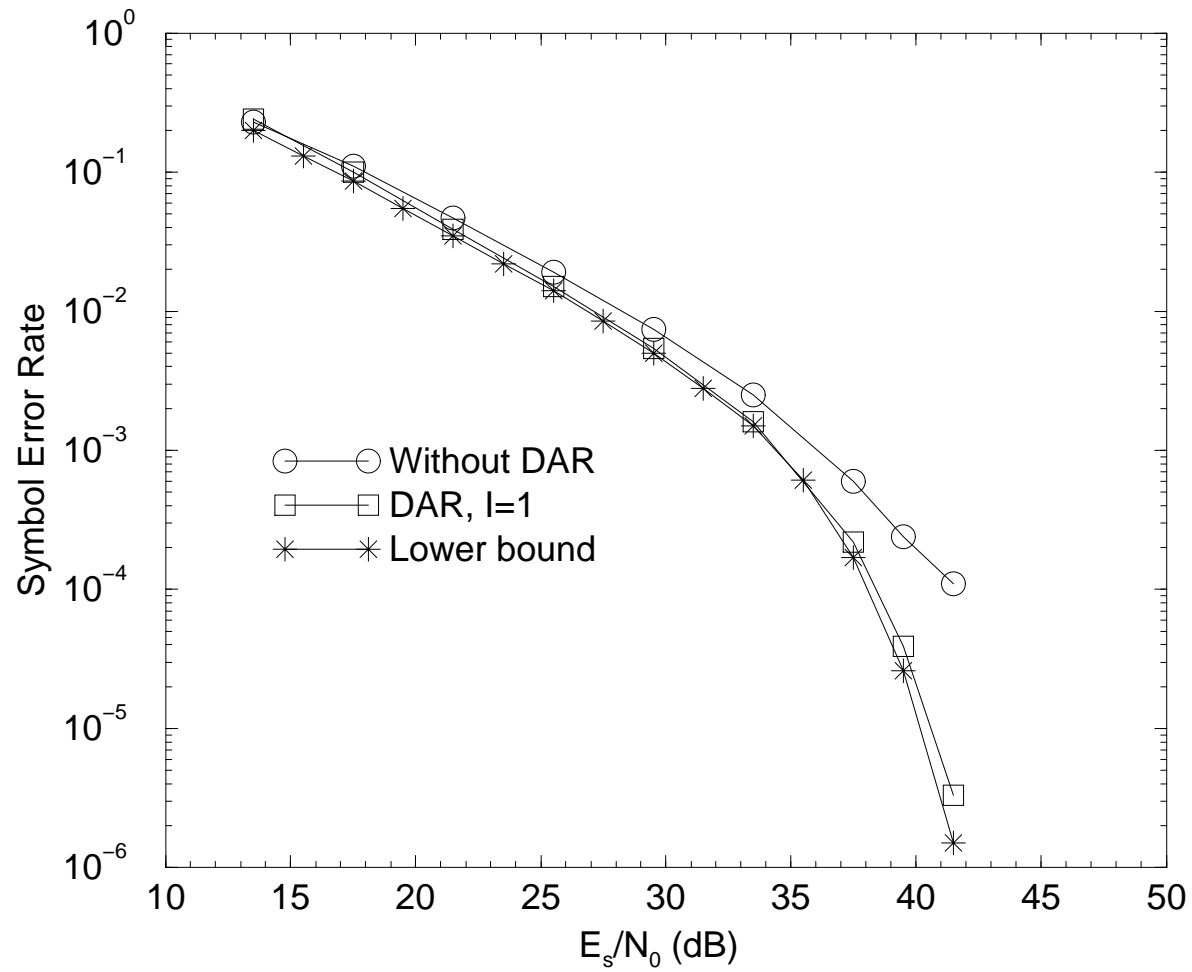
tap	delay (μs)	tap coeff.
h_0	0.0	0.405
h_1	0.2	0.541
h_2	1.0	0.383
h_3	1.6	0.307
h_4	5.0	0.430
h_5	6.6	0.342

Simulation Results



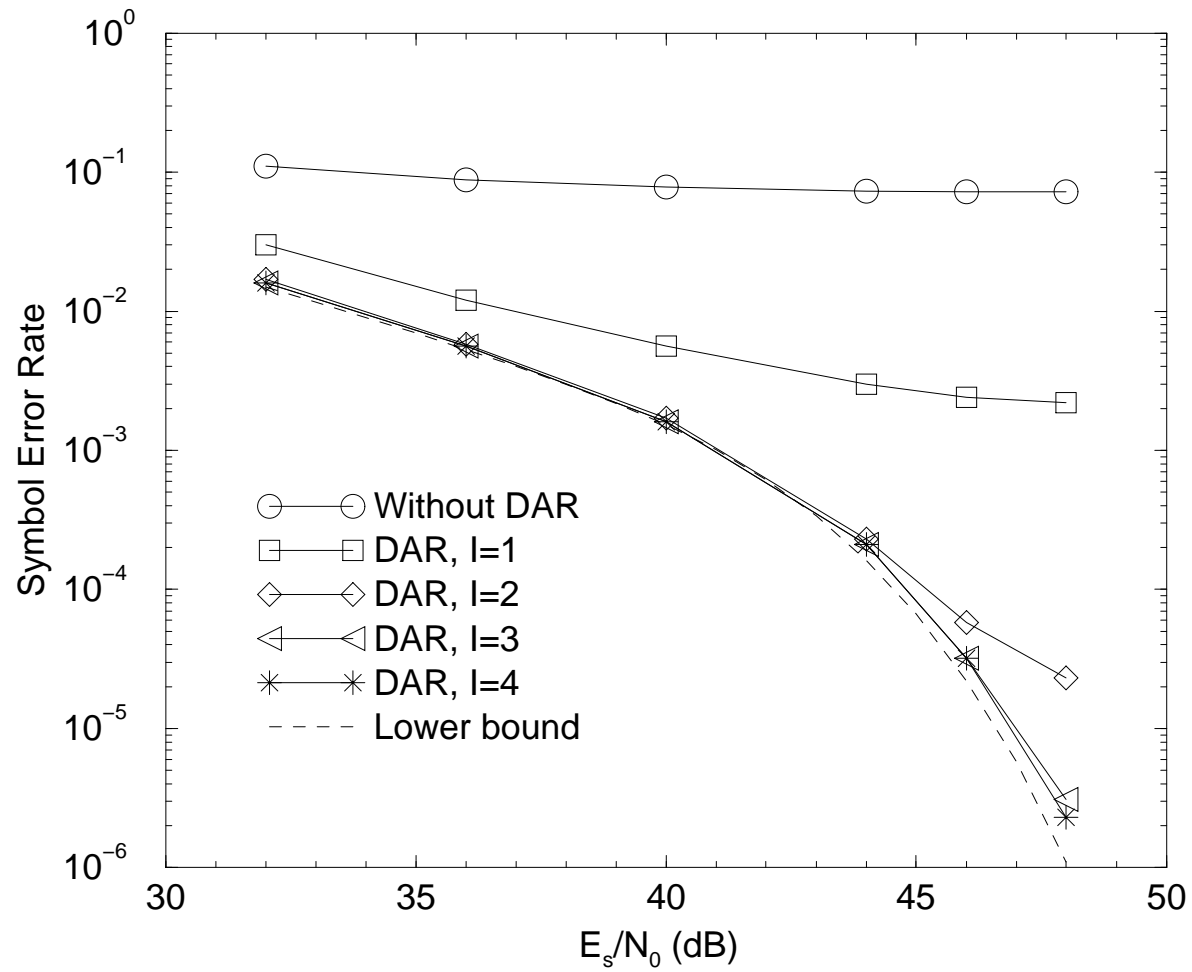
On AWGN channel with 64-QAM; CR=5 dB.
The lower bound is achieved when clipping is not used.

Simulation Results



On static ISI channel with 16-QAM; CR=4 dB, $N = 1024$.

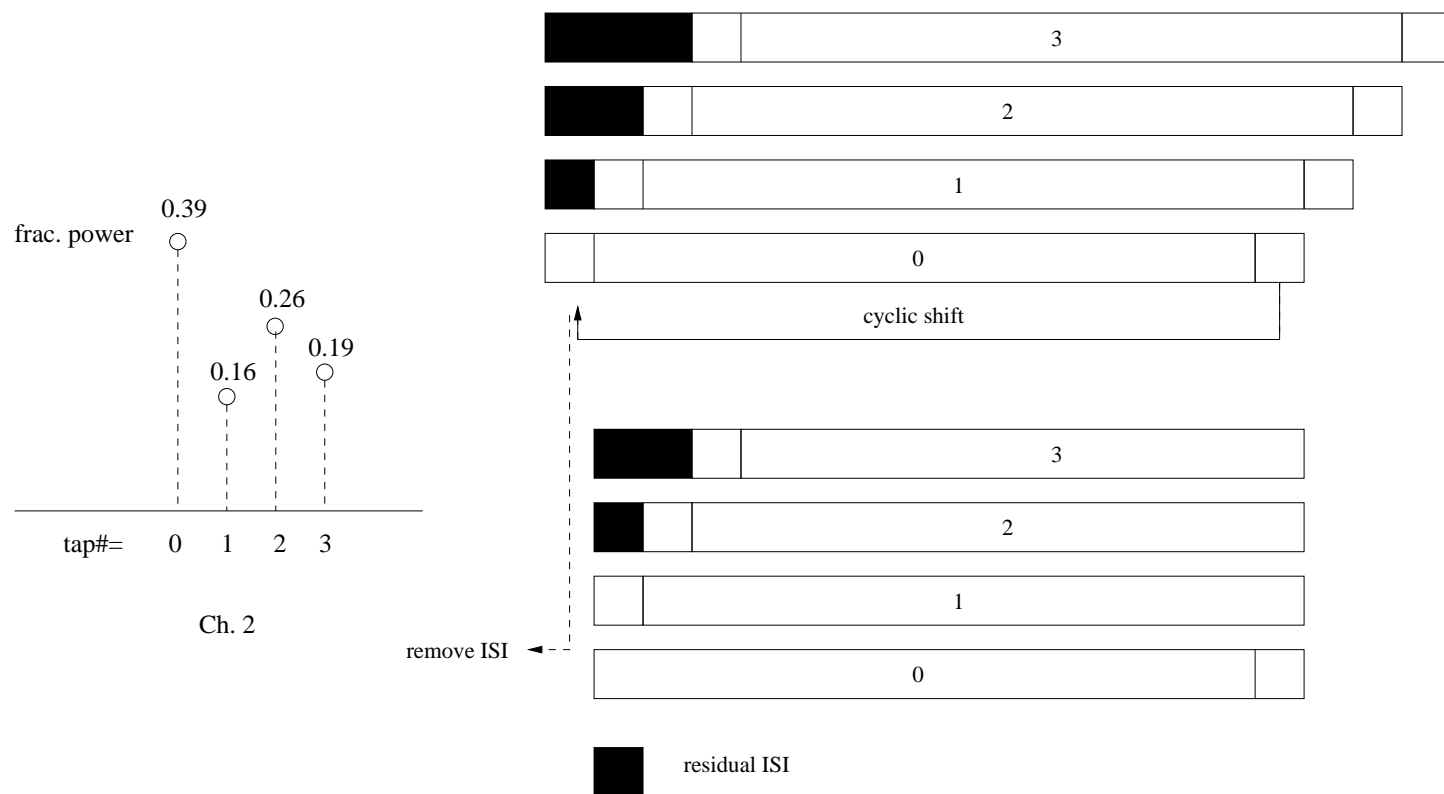
Simulation Results



On static ISI channel with 64-QAM; CR=4 dB, $N = 1024$.

Mitigation of ISI Using Guard Interval

- Assume a guard interval of length $G = 1$ and a channel of length $L = 3$. In this case, there will be residual ISI.



Residual ISI Cancellation (RISIC)

- **RISIC algorithm is carried out by executing two steps, *tail cancellation* and *cyclic reconstruction*.**

1. **Decisions on transmitted symbols $\{\hat{x}_{i-1,n}\}_{n=0}^{N-1}$ from block $i-1$ are obtained for use in tail cancellation. These symbols are converted back to the time domain using an IFFT giving $\{\hat{X}_{i-1,k}\}_{k=0}^{N-1}$.**
2. **For block i , we perform tail cancellation by calculating residual ISI and subtract it from $R_{i,k}$ as**

$$\tilde{R}_{i,k}^{(0)} = R_{i,k} - \sum_{m=G+1}^L h_m \hat{X}_{i-1,(k-m+G)_N} (1 - u(k - m + G)) \quad 0 \leq k \leq N - 1$$

3. **$\{\tilde{R}_{i,k}^{(0)}\}_{k=0}^{N-1}$ obtained in Step 2 are converted to frequency domain using FFT and decisions are made. Afterwards, decisions are converted back to time domain to give $\{\hat{X}_{i,k}^{(0)}\}_{k=0}^{N-1}$.**

RISIC Algorithm

4. Next we perform cyclic reconstruction by forming

$$\tilde{R}_{i,k}^{(I)} = \tilde{R}_{i,k}^{(0)} + \sum_{m=G+1}^L h_m \hat{X}_{i,(k-m)_N}^{(I-1)} (1 - u(k - m + G)), \quad 0 \leq k \leq N - 1$$

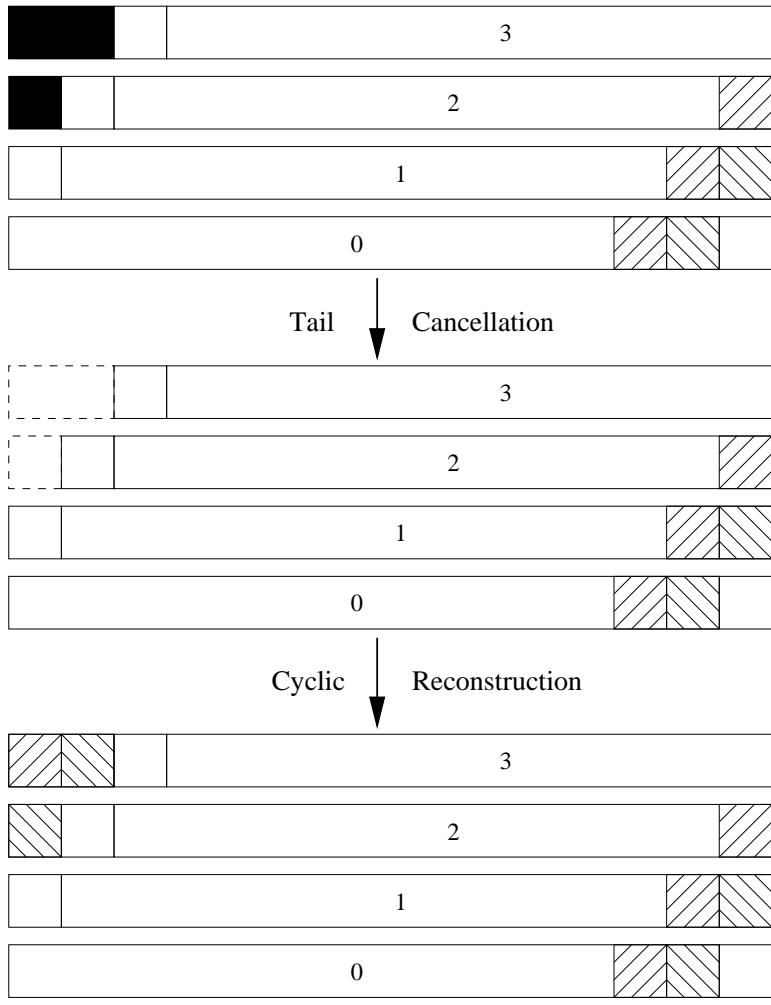
where I represents an iteration number with an initial value of $I = 1$.

5. $\{\tilde{R}_{i,k}^{(I)}\}_{k=0}^{N-1}$ are converted to frequency domain and decisions are made yielding $\{\hat{x}_{i,n}^{(I)}\}_{n=0}^{N-1}$. This completes the I th iteration in RISIC algorithm.

6. To continue iterations, convert $\{\hat{x}_{i,n}^{(I)}\}_{n=0}^{N-1}$ to $\{\hat{X}_{i,k}^{(I)}\}_{k=0}^{N-1}$ and repeat Steps 4 – 6 with $I \leftarrow I + 1$.

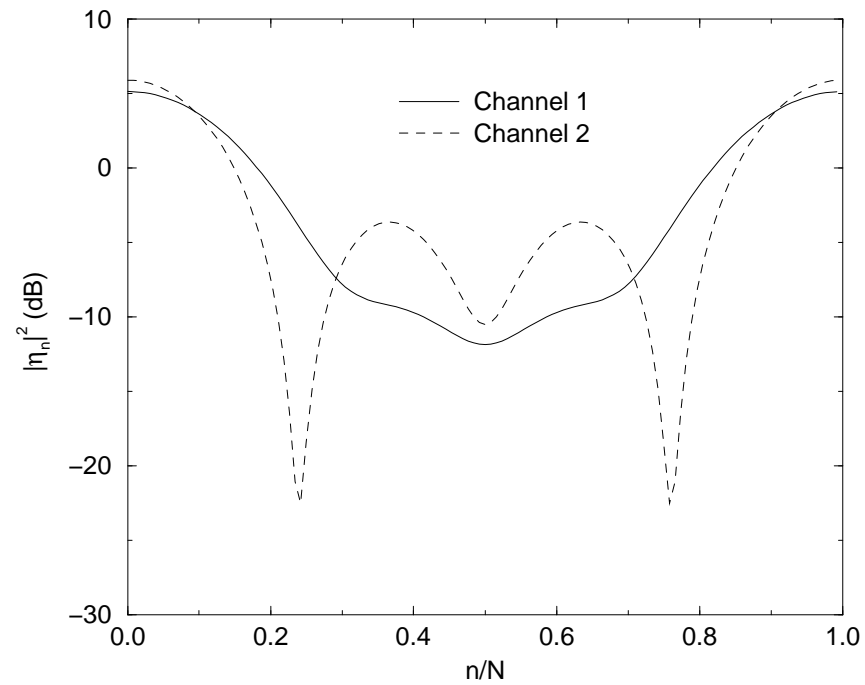
7. End of RISIC algorithm for block i .

RISIC Algorithm



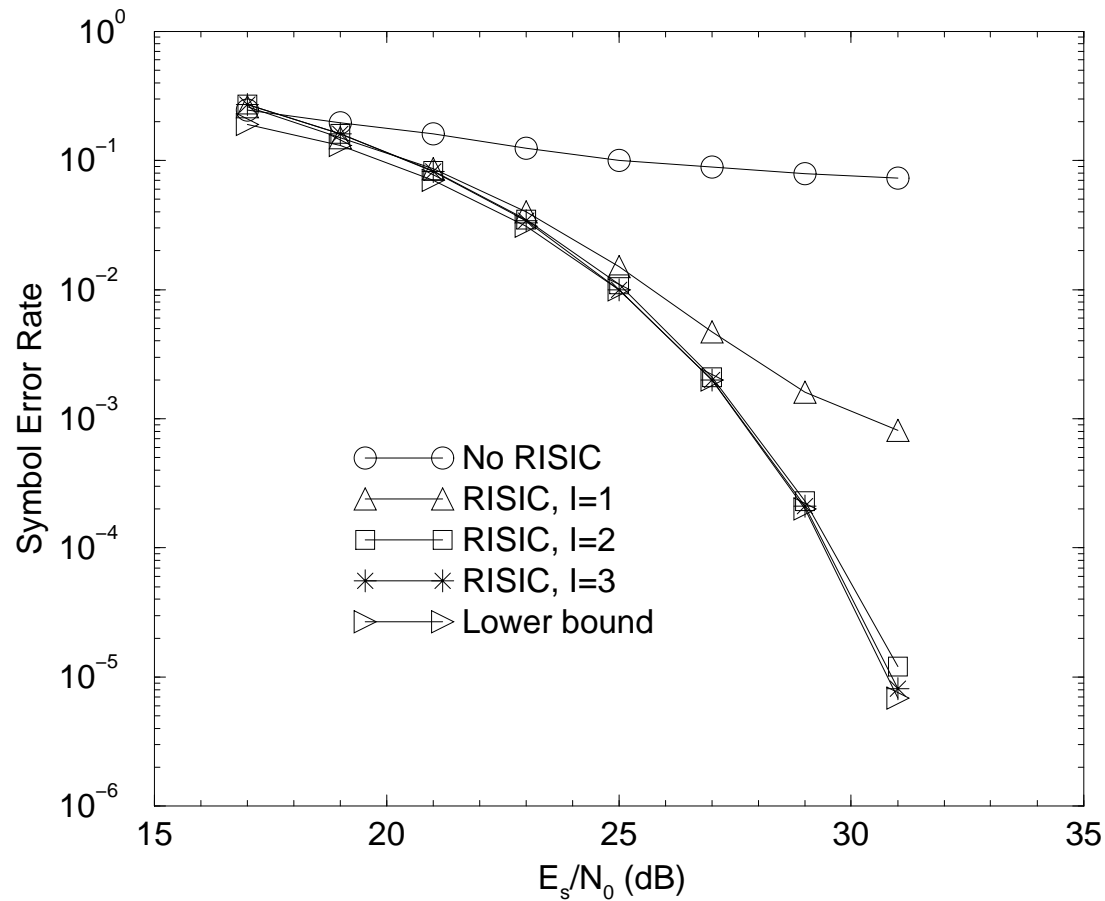
Static Channels ($N = 128$)

tap #	delay (μs)	Frac. Power (Ch. 1)	Frac. Power (Ch. 2)
0	0.0	0.15	0.39
1	0.2	0.65	0.16
2	0.4	0.15	0.26
3	0.6	0.05	0.19



16-QAM on Static Channel; 20 Mbps

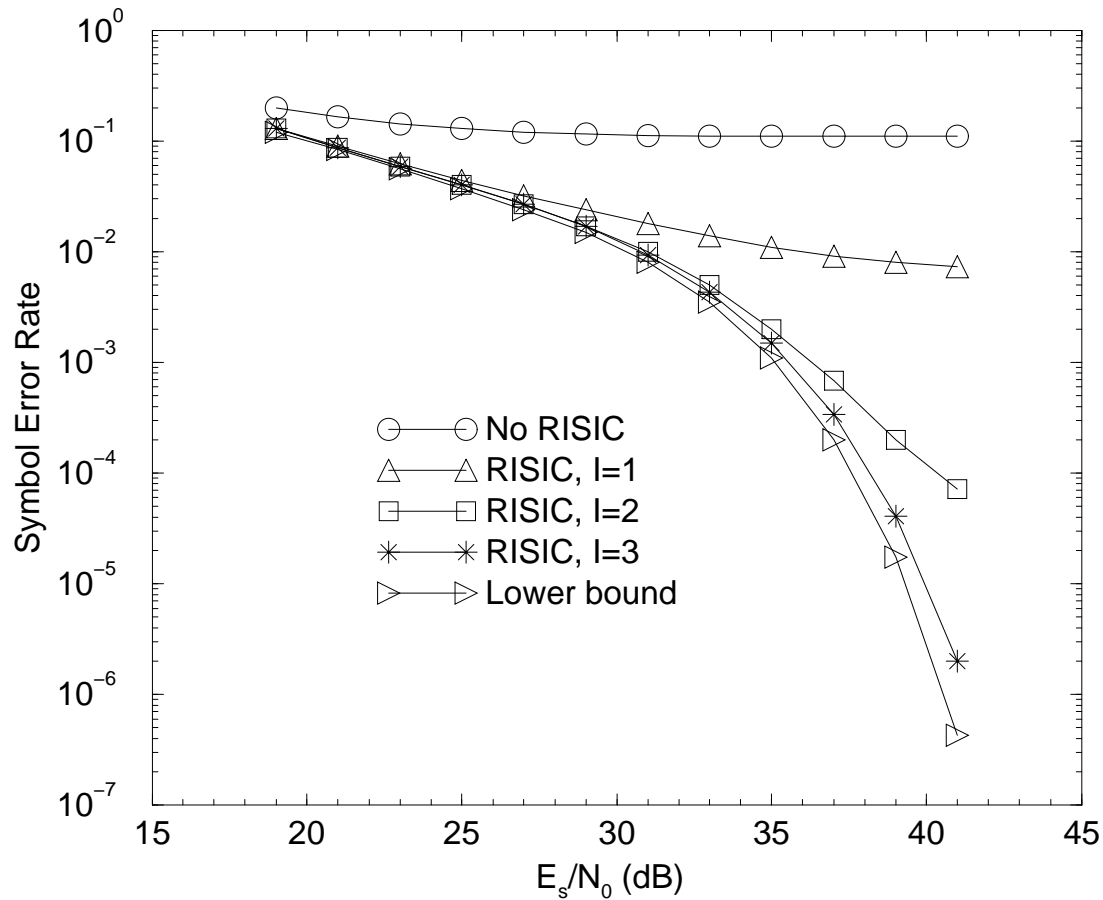
Ch.1, $G = 0$, $N = 128$



RISIC gives a severe error floor. RISIC recovers with $I = 2$ iterations.

16-QAM on Static Channel; 20 Mbps

Ch.2, $G = 0$, $N = 128$



RISIC gives a severe error floor. RISIC recovers with $I = 3$ iterations.